

**Single Output** Half-brick

# MILITARY COTS DC-DC CONVERTER

34-75V Continuous Input Transient Input

34-100V

Output

60A Output 92% @ 30A / 89% @ 60A

**Efficiency** 

Operation: -55°C to +100°C

The MilQor series of Mil-COTS DC-DC converters brings SynQor's field proven high-efficiency synchronous rectification technology to the Military/Aerospace industry. SynQor's ruggedized encased packaging approach ensures survivability in demanding environments. Compatible with the industry standard format, these converters operate at a fixed frequency, and follow conservative component derating guidelines. They are designed and manufactured to comply with a wide range of military standards.

# **Mil-COTS**



# **Safety Features**

- 2250V, 30 MΩ input-to-output isolation
- Certified 60950-1 requirement for basic insulation (see Standards and Qualifications page)

#### **Mechanical Features**

- Industry standard half-brick pin-out configuration
- 2.49" x 2.39" x 0.51" Size:

(63.1 x 60.6 x 13.0 mm)

- Total weight: 4.7 oz. (134 g)
- Flanged baseplate version available

#### **Protection Features**

- Input under-voltage lockout
- Output current limit and short circuit protection
- Active back bias limit
- Output over-voltage protection
- Thermal shutdown

# **Specification Compliance**

MCOTS series converters (with an MCOTS filter) are designed to meet:

- MIL-HDBK-704 (A-F)
- RTCA/DO-160E Section 16
- MIL-STD-1275 (B,D)
- DEF-STAN 61-5 (Part 6)/(5 or 6)
- MIL-STD-461 (C, D, E, F)

Designed and Manufactured in the USA

#### **Operational Features**

- · High efficiency, 89% at full rated load current
- Operating input voltage range: 34-75V
- Fixed frequency switching provides predictable EMI
- No minimum load requirement

#### **Control Features**

- On/Off control referenced to input return
- Remote sense for the output voltage
- Output voltage trim range of +10%, -20%

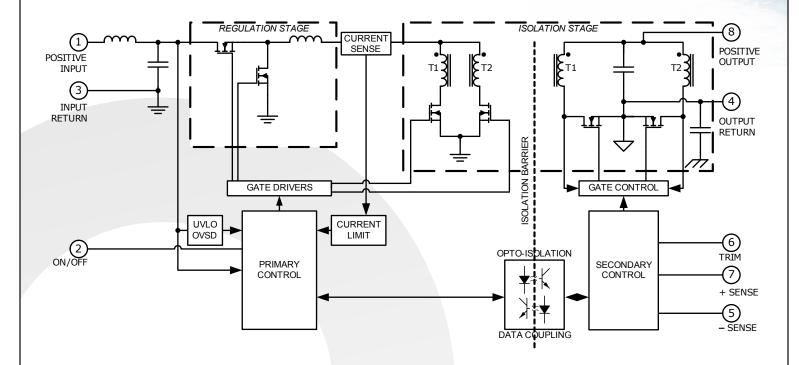
# **Screening/Qualification**

- AS9100 and ISO 9001:2008 certified facility
- Qualified to MIL-STD-810
- Available with S-Grade or M-Grade screening
- Temperature cycling per MIL-STD-883, Method 1010, Condition B, 10 cycles
- Burn-In at 100C baseplate temperature
- Final visual inspection per MIL-STD-2008
- Full component traceability

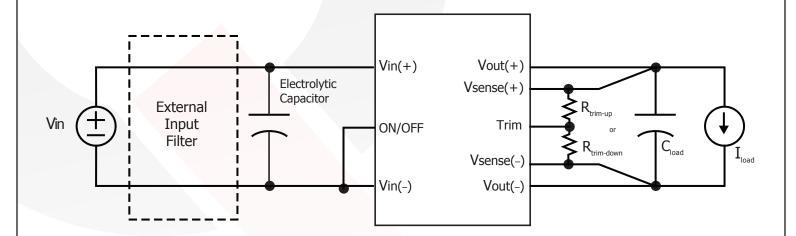
Output: 3.3V

**Current: 60A** 

# **BLOCK DIAGRAM**



# **TYPICAL CONNECTION DIAGRAM**



Output: 3.3V Current: 60A

# MCOTS-C-48-3R3-HP ELECTRICAL CHARACTERISTICS

Tb = 25 °C, Vin = 48Vdc unless otherwise noted; full operating temperature range is -55 °C to +100 °C baseplate temperature with appropriate power derating. Specifications subject to change without notice.

Parameter	Min.	Тур.	Max.	Units	Notes & Conditions
ABSOLUTE MAXIMUM RATINGS					
Input Voltage					
Non-Operating	-1		100	V	Continuous
Operating			75	V	Continuous
Operating Transient Protection			100	V	100 ms transient, square wave
Isolation Voltage					
Input to Output			2250	V dc	
Input to Base-Plate			2250	V dc	
Output to Base-Plate			2250	V dc	
Operating Temperature	-55		100	°C	Baseplate temperature
Storage Temperature	-65		135	°C	Buschiete temperature
Voltage at ON/OFF input pin	-2		18	V	
INPUT CHARACTERISTICS	2		10	V	
Operating Input Voltage Range	34	48	75	V	100V transient for 100 ms
Input Under-Voltage Lockout	JT	iO	,5	V	100 V (I di i sicriti i or 100 i i i s
Turn-On Voltage Threshold	32.7	33.1	33.5	V	
Turn-Off Voltage Threshold	30.4	30.8	31.2	V	
Lockout Voltage Hysteresis		2.3		V	
Recommended External Input Capacitance		100		μF	Typical ESR 0.1-0.2 Ω
nput Filter Component Values (L\C)		3.4\4.0		μΗ\μF	Internal values; see Figure E
Maximum Input Current		,	8.6	A	Vin min; trim up; in current limit
No-Load Input Current		60	100	mA	The time, come up, in content in the
Disabled Input Current		1	2	mA	
Response to Input Transient		0.24		V	See Figure 12
Input Terminal Ripple Current		200		mA	RMS
Recommended Input Fuse		200	20	A	Fast acting external fuse recommended
OUTPUT CHARACTERISTICS			20		Tast acting external ruse recommended
Output Voltage Set Point	3.267	3.3	3,333	V	
Output Voltage Regulation	3.207	3.3	3.333	V	
Over Line		±0.1	±0.3	%	
Over Load		±0.1	±0.3	%	
	40 F	±0.1			
Over Temperature	-49.5 2 219		49.5	mV	Over sample line lead temperature 9 life
Total Output Voltage Range	3.218		3.383	V	Over sample, line, load, temperature & life
Output Voltage Ripple and Noise Peak-to-Peak	0	80	160	l mV	20 MHz bandwidth; see Note 1 Full load
RMS	U	20	40	mV	Full load
Operating Output Current Range	0	20	60	A	Subject to thermal derating
Output DC Current-Limit Inception	66	72	78	A	Output voltage 10% Low
Output DC Current-Limit Inception  Output DC Current-Limit Shutdown Voltage	00	1.4	70	V	Output voltage 10 /0 LOW
Back-Drive Current Limit while Enabled		1.2		A	Negative current drawn from output
Back-Drive Current Limit while Disabled		1.2		mA	Negative current drawn from output
Maximum Output Capacitance		1	10000		Vout nominal at full load (resistive load)
			10000	μF	vout nominal at full load (resistive load)
Output Voltage during Load Current Transient		90		pa\/	E00/ to 750/ to 500/ Tout many
Step Change in Output Current (0.1 A/µs)		80		mV	50% to 75% to 50% Iout max
Settling Time	20	400	10	μs	To within 1% Vout nom
Output Voltage Trim Range	-20		10	%	Across Pins 8&4; Figure C; See Note 2
Output Voltage Remote Sense Range	2.0	4.0	10	%	Across Pins 8&4
Output Over-Voltage Protection	3.9	4.0	4.2	V	Over full temp range
FFICIENCY		00		C.	lc 5: 4.5 (5: :
100% Load		89		%	See Figure 1 for efficiency curve
50% Load		92		%	See Figure 1 for efficiency curve

Note 1: Output is terminated with 1 µF ceramic and 15 µF low-ESR tantalum capacitors. For applications requiring reduced output voltage ripple and noise, consult SynQor applications support (e-mail: mqnbofae@synqor.com)

Note 2: Trim-up range is limited below 10% at low line and full load.

Product # MCOTS-C-48-3R3-HP Phone 1-888-567-9596 www.synqor.com Doc.# 005-0005375 Rev. E 09/02/14 Page 3

Output: 3.3V **Current: 60A** 

# MCOTS-C-48-3R3-HP ELECTRICAL CHARACTERISTICS

Tb = 25 °C, Vin = 48Vdc unless otherwise noted; full operating temperature range is -55 °C to +100 °C baseplate temperature with appropriate power derating. Specifications subject to change without notice.

Turn-On Transient Turn-On Transient Turn-On Time Start-Up Inhibit Time Output Voltage Overshoot  ISOLATION CHARACTERISTICS Isolation Voltage (dielectric strength) Isolation Resistance Isolation Capacitance (input to output)  TEMPERATURE LIMITS FOR POWER DERATING CURVE Semiconductor Junction Temperature Board Temperature Transformer Temperature Maximum Baseplate Temperature, Tb	<b>Typ.</b>			
Turn-On Time Start-Up Inhibit Time Output Voltage Overshoot  ISOLATION CHARACTERISTICS Isolation Voltage (dielectric strength) Isolation Resistance Isolation Capacitance (input to output)  TEMPERATURE LIMITS FOR POWER DERATING CURVE Semiconductor Junction Temperature Board Temperature Transformer Temperature	10			
Start-Up Inhibit Time Output Voltage Overshoot  ISOLATION CHARACTERISTICS Isolation Voltage (dielectric strength) Isolation Resistance Isolation Capacitance (input to output)  TEMPERATURE LIMITS FOR POWER DERATING CURVE Semiconductor Junction Temperature Board Temperature Transformer Temperature	10			
Output Voltage Overshoot  ISOLATION CHARACTERISTICS  Isolation Voltage (dielectric strength)  Isolation Resistance Isolation Capacitance (input to output)  TEMPERATURE LIMITS FOR POWER DERATING CURVE Semiconductor Junction Temperature Board Temperature Transformer Temperature			ms	Full load, Vout=90% nom.
ISOLATION CHARACTERISTICS Isolation Voltage (dielectric strength) Isolation Resistance Isolation Capacitance (input to output) TEMPERATURE LIMITS FOR POWER DERATING CURVE Semiconductor Junction Temperature Board Temperature Transformer Temperature	200	220	ms	Figure F
Isolation Voltage (dielectric strength) Isolation Resistance Isolation Capacitance (input to output) TEMPERATURE LIMITS FOR POWER DERATING CURVE Semiconductor Junction Temperature Board Temperature Transformer Temperature	0		%	Maximum Output Capacitance
Isolation Resistance Isolation Capacitance (input to output) TEMPERATURE LIMITS FOR POWER DERATING CURVE Semiconductor Junction Temperature Board Temperature Transformer Temperature				
Isolation Capacitance (input to output)  TEMPERATURE LIMITS FOR POWER DERATING CURVE Semiconductor Junction Temperature Board Temperature Transformer Temperature				See Absolute Maximum Ratings
TEMPERATURE LIMITS FOR POWER DERATING CURVE Semiconductor Junction Temperature Board Temperature Transformer Temperature	30		MΩ	
Semiconductor Junction Temperature Board Temperature Transformer Temperature	1000		pF	See Note 1
Board Temperature Transformer Temperature	S			
Transformer Temperature		125	°C	Package rated to 150 °C
		125	°C	UL rated max operating temp 130 °C
Maximum Baseplate Temperature, Tb		125	°C	
		100	°C	
FEATURE CHARACTERISTICS				
Switching Frequency 255	275	295	kHz	Isolation stage switching freq. is half this
ON/OFF Control				
Off-State Voltage 2.4		1.8	V	
On-State Voltage -2		0.8		
ON/OFF Control				Application notes Figures A & B
Pull-Up Voltage	5		V	
Pull-Up Resistance	50		kΩ	
Over-Temperature Shutdown OTP Trip Point	125		°C	Average PCB Temperature
Over-Temperature Shutdown Restart Hysteresis	10		°C	
RELIABILITY CHARACTERISTICS				
Calculated MTBF per MIL-HDBK-217F	4.1		10 <sup>6</sup> Hrs.	Ground Benign, 70°C Tb
Calculated MTBF per MIL-HDBK-217F	11.2		10 1113.	Ground Berlight, 70 C 15

Note 1: Higher values of isolation capacitance can be added external to the module.

#### STANDARDS COMPLIANCE

**Notes & Conditions Parameter** 

**STANDARDS COMPLIANCE** 

UL 60950-1/R:2011-12 **Basic Insulation** 

CAN/CSA-C22.2 No. 60950-1/A1:2011

EN60950-1/A12:2011

Note: An external input fuse must always be used to meet these safety requirements.

Contact SynQor for official safety certificates on new releases or download from the SynQor website.

Product # MCOTS-C-48-3R3-HP Phone 1-888-567-9596 www.synqor.com Doc.# 005-0005375 Rev. E 09/02/14 Page 4



Output: 3.3V Current: 60A

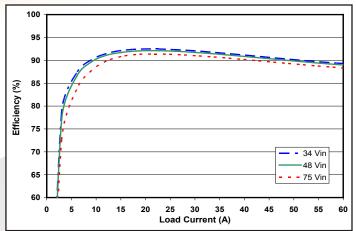


Figure 1: Efficiency at nominal output voltage vs. load current for minimum, nominal, and maximum input voltage at 25°C.

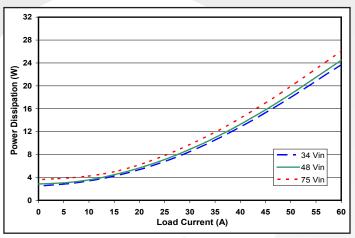


Figure 3: Power dissipation at nominal output voltage vs. load current for minimum, nominal, and maximum input voltage at  $T_{CASE}$ =25°C.

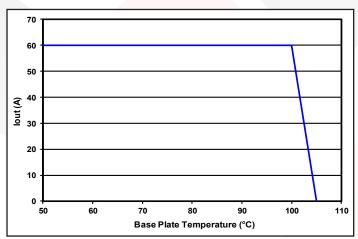


Figure 5: Maximum output current vs. base plate temperature (nominal input voltage).

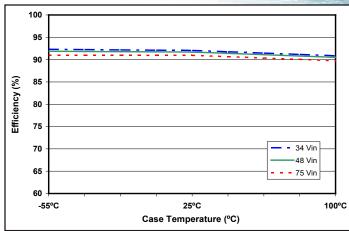


Figure 2: Efficiency at nominal output voltage and 60% rated power vs. case temperature for minimum, nominal, and maximum input voltage.

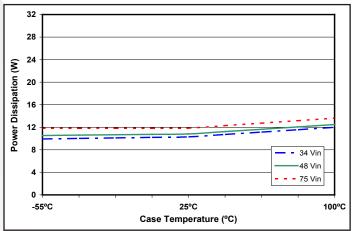


Figure 4: Power dissipation at nominal output voltage and 60% rated power vs. case temperature for minimum, nominal, and maximum input voltage.

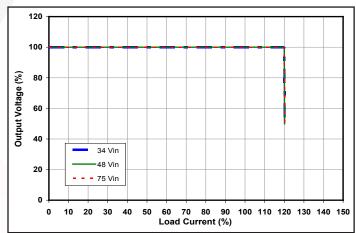


Figure 6: Output voltage vs. load current showing typical current limit curves. See Current Limit section in the Application Notes.



Output: 3.3V Current: 60A

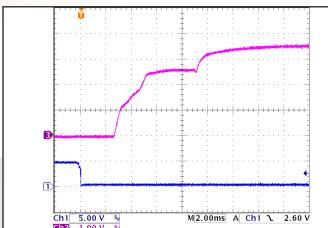
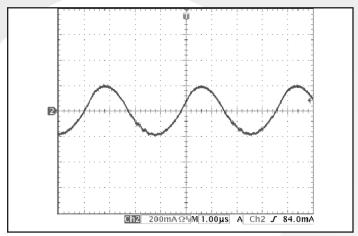


Figure 7: Typical startup waveform. Input voltage pre-applied, ON/OFF Pin. Ch 1: ON/OFF Pin (5V/div). Ch 3: Vout (1V/div).



**Figure 9:** Input terminal current ripple, i<sub>C</sub>, at full rated output current and nominal input voltage with SynQor MCOTS filter module (200 mA/div). Bandwidth: 20MHz. See Figure 17.

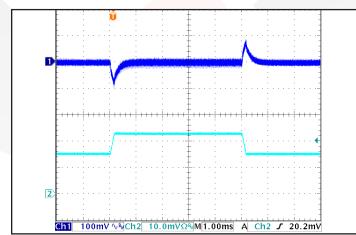


Figure 11: Output voltage response to step-change in load current (50%-75%-50% of Iout(max);  $dI/dt = 0.1 A/\mu s$ ). Load cap:  $1 \mu F$  ceramic and  $15 \mu F$  tantalum capacitors. Ch 1: Vout (100mV/div), Ch 2: Iout (20A/div).

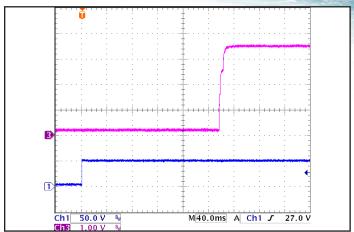
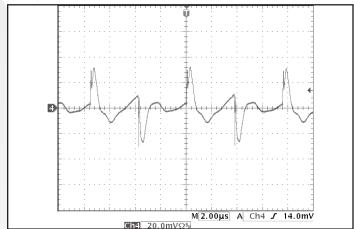


Figure 8: Turn-on transient at full resistive load and zero output capacitance initiated by Vin. ON/OFF Pin previously low. Ch 1: Vin (50V/div). Ch 3: Vout (1V/div).



**Figure 10:** Output voltage ripple, Vout, at nominal input voltage and rated load current (20 mV/div). Load capacitance:  $1\mu$ F ceramic capacitor and  $10\mu$ F tantalum capacitor. Bandwidth: 10 MHz. See Figure 17.

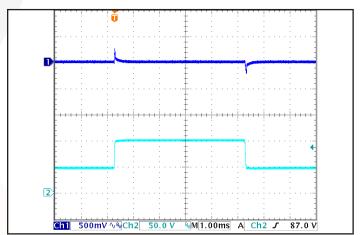


Figure 12: Output voltage response to step-change in input voltage (250V/ms). Load cap: 100 μF electrolytic output capacitance. Ch 1: Vout (500mV/div), Ch 2: Vin (50V/div).



Output: 3.3V **Current: 60A** 

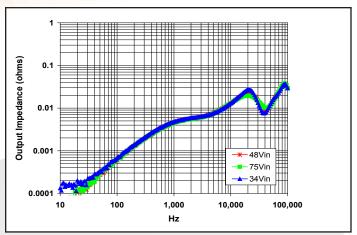
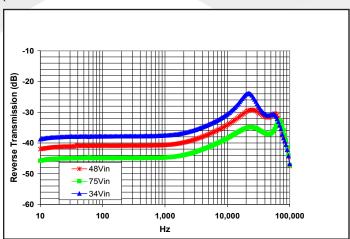


Figure 13: Magnitude of incremental output impedance  $(Z_{out} = v_{out})$ iout) for minimum, nominal, and maximum input voltage at full rated



**Figure 15:** Magnitude of incremental reverse transmission (RT =  $i_{in}$ / iout) for minimum, nominal, and maximum input voltage at full rated power.

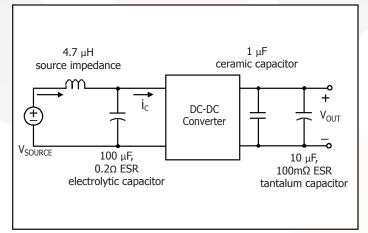


Figure 17: Test set-up diagram showing measurement points for Input Terminal Ripple Current (Figure 9) and Output Voltage Ripple (Figure 10).

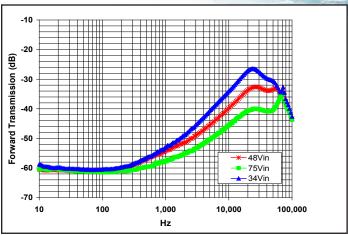


Figure 14: Magnitude of incremental forward transmission (FT =  $v_{out}$ ) vin) for minimum, nominal, and maximum input voltage at full rated power.

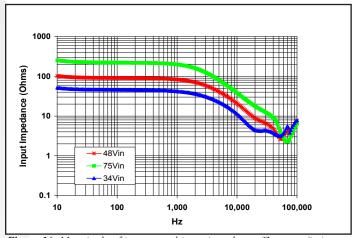


Figure 16: Magnitude of incremental input impedance  $(Z_{in} = v_{in}/i_{in})$ for minimum, nominal, and maximum input voltage at full rated power.

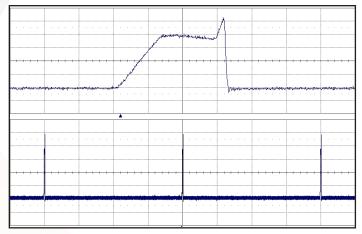


Figure 18: Load current (20A/div) as a function of time (0.5ms/div and 50ms/ div) when the converter attempts to turn on into a 10 m $\Omega$  short circuit.

# BASIC OPERATION AND FEATURES

This Mil-COTS converter series uses a two-stage power conversion topology. The first stage is a buck-converter that keeps the output voltage constant over variations in line, load, and temperature. The second stage uses a transformer to provide the functions of input/output isolation and voltage step-up or step-down to achieve the output voltage required.

Both the first stage and the second stage switch at a fixed frequency for predictable EMI performance. Rectification of the transformer's output is accomplished with synchronous rectifiers. These devices, which are MOSFETs with a very low on-state resistance, dissipate far less energy than Schottky diodes. This is the primary reason that the converter has such high efficiency, even at very low output voltages and very high output currents.

These converter are offered totally encased to withstand harsh environments and thermally demanding applications. Dissipation throughout the converter is so low that it does not require a heatsink for operation in many applications; however, adding a heatsink provides improved thermal derating performance in extreme situations.

This series of converters use the industry standard footprint and pin-out configuration.

# CONTROL FEATURES

**REMOTE ON/OFF (Pin 2)**: The ON/OFF input, Pin 2, permits the user to control when the converter is on or off. This input is referenced to the return terminal of the input bus, Vin(-). The ON/OFF signal is active low (meaning that a low turns the converter on). Figure A details four possible circuits for driving the ON/OFF pin. Figure B is a detailed look of the internal ON/ OFF circuitry.

**REMOTE SENSE(\pm)** (Pins 7 and 5): The SENSE( $\pm$ ) inputs correct for voltage drops along the conductors that connect the converter's output pins to the load.

Pin 7 should be connected to Vout(+) and Pin 5 should be connected to Vout(-) at the point on the board where regulation is desired. A remote connection at the load can adjust for a voltage drop only as large as that specified in this datasheet, that is

$$[Vout(+) - Vout(-)] - [Vsense(+) - Vsense(-)] \le$$
  
Sense Range % x Vout

Pins 7 and 5 must be connected for proper regulation of the output voltage. If these connections are not made, the converter will deliver an output voltage that is slightly higher than its specified value.

Note: the output over-voltage protection circuit senses the voltage across the output (pins 8 and 4) to determine when it should trigger, not the voltage across the converter's sense leads (pins 7 and 5). Therefore, the resistive drop on the board should be small enough so that output OVP does not trigger, even during load transients.

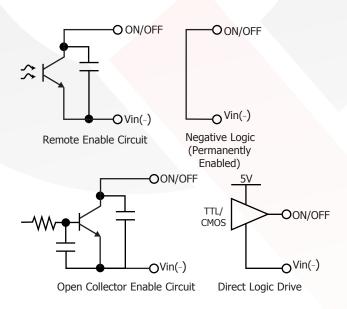


Figure A: Various circuits for driving the ON/OFF pin.

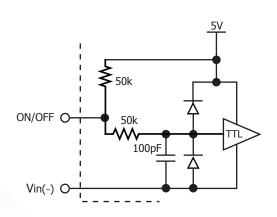


Figure B: Internal ON/OFF pin circuitry

**OUTPUT VOLTAGE TRIM (Pin 6)**: The TRIM input permits the user to adjust the output voltage across the sense leads up or down according to the trim range specifications.

To decrease the output voltage, the user should connect a resistor between Pin 6 and Pin 5 (SENSE(-) input). For a desired decrease of the nominal output voltage, the value of the resistor should be

$$R_{\text{trim-down}} = \left(\frac{511}{\Lambda\%}\right) - 10.22 \text{ (k}\Omega)$$

where

$$\Delta\% = \left| \frac{\text{Vnominal} - \text{Vdesired}}{\text{Vnominal}} \right| \times 100\%$$

To increase the output voltage, the user should connect a resistor between Pin 6 and Pin 7 (SENSE(+) input). For a desired increase of the nominal output voltage, the value of the resistor should be

R<sub>trim-up</sub> = 
$$\left(\frac{5.11 V_{out} \times (100 + \Delta\%)}{1.225 \Delta\%} - \frac{511}{\Delta\%} - 10.22\right)$$
 (kΩ)  
where  $V_{out}$  = Nominal Output Voltage

Trim graphs show the relationship between the trim resistor value and Rtrim-up and Rtrim-down, showing the total range the output voltage can be trimmed up or down.

<u>Note</u>: the TRIM feature does not affect the voltage at which the output over-voltage protection circuit is triggered. Trimming the output voltage too high may cause the over-voltage protection circuit to engage, particularly during transients.

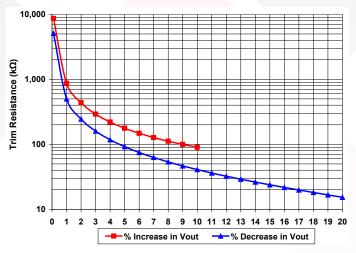


Figure C: Trim graph for trim-up, trim down.

It is not necessary for the user to add capacitance at the Trim pin. The node is internally bypassed to eliminate noise.

**Total DC Variation of V**<sub>our</sub>: For the converter to meet its full specifications, the maximum variation of the dc value of  $V_{our}$ , due to both trimming and remote load voltage drops, should not be greater than that specified for the output voltage trim range.

# PROTECTION FEATURES

**Input Under-Voltage Lockout**: The converter is designed to turn off when the input voltage is too low, helping avoid an input system instability problem, described in more detail in the application note titled "Input System Instability" on our website. The lockout circuitry is a comparator with dc hysteresis. When the input voltage is rising, it must exceed the typical Turn-On Voltage Threshold value (listed on the specifications page) before the converter will turn on. Once the converter is on, the input voltage must fall below the typical Turn-Off Voltage Threshold value before the converter will turn off.

**Output Over-Voltage Limit**: If the voltage across the output pins exceeds the Output Over-Voltage Protection threshold, the converter will immediately stop switching. This prevents damage to the load circuit due to 1) excessive series resistance in output current path from converter output pins to sense point, 2) a release of a short-circuit condition, or 3) a release of a current limit condition. Load capacitance determines exactly how high the output voltage will rise in response to these conditions. After 200 ms the converter will automatically restart.

**Over-Temperature Shutdown**: A temperature sensor on the converter senses the average temperature of the module. The thermal shutdown circuit is designed to turn the converter off when the temperature at the sensed location reaches the Over-Temperature Shutdown value. It will allow the converter to turn on again when the temperature of the sensed location falls by the amount of the Over-Temperature Shutdown Restart Hysteresis value.

**Output Current Limit:** The maximum current limit remains constant as the output voltage drops. However, once the impedance of the load across the output is small enough to make the output voltage drop below the specified Output DC Current-Limit Shutdown Voltage, the converter turns off.

The converter then enters a "hiccup" mode where it repeatedly turns on and off at a 5 Hz (nominal) frequency with 5% duty cycle until the short circuit condition is removed. This prevents excessive heating of the converter or the load board.

# APPLICATION CONSIDERATIONS

**Input System Instability:** This condition can occur because any dc-dc converter appears incrementally as a negative resistance load. A detailed application note titled "Input System Instability" is available on the SynQor website which provides an understanding of why this instability arises, and shows the preferred solution for correcting it.

**Application Circuits:** Figure D provides a typical circuit diagram which details the input filtering and voltage trimming.

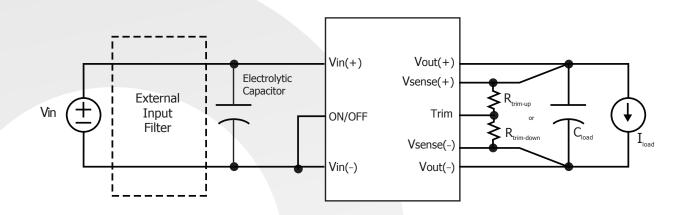


Figure D: Typical application circuit (negative logic unit, permanently enabled).

**Input Filtering and External Capacitance:** Figure E provides a diagram showing the internal input filter components. This filter dramatically reduces input terminal ripple current, which otherwise could exceed the rating of the converter's external electrolytic input capacitor. The recommended external input capacitance

is specified in the Input Characteristics section on the Electrical Characteristics page. More detailed information is available in the application note titled "EMI Characteristics" on the SynQor website.

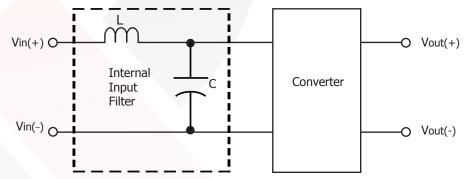


Figure E: Internal Input Filter Diagram

**Startup Inhibit Period:** The Startup Inhibit Period ensures that the converter will remain off for approximately 200 ms when it is shut down for any reason. When an output short is present, this generates a 5 Hz "hiccup" mode, which prevents the converter from overheating. In all, there are seven ways that the converter can be shut down, initiating a Startup Inhibit Period:

- Input Under-Voltage Lockout
- Input Over-Voltage Lockout
- Output Over-Voltage Protection
- Over Temperature Shutdown
- Current Limit
- Short Circuit Protection
- Turned off by the ON/OFF input

Figure F shows three turn-on scenarios, where a Startup Inhibit Period is initiated at  $t_0$ ,  $t_1$ , and  $t_2$ :

Before time  $t_0$ , when the input voltage is below the UVL threshold, the unit is disabled by the Input Under-Voltage Lockout feature. When the input voltage rises above the UVL threshold, the Input Under-Voltage lockout is released, and a Startup Inhibit Period is initiated. At the end of this delay, the ON/OFF pin is evaluated, and since it is active, the unit turns on.

At time t<sub>1</sub>, the unit is disabled by the ON/OFF pin, and it cannot be enabled again until the Startup Inhibit Period has elapsed.

When the ON/OFF pin goes high after  $t_2$ , the Startup Inhibit Period has elapsed, and the output turns on within the typical Turn-On Time.

**Thermal Considerations:** The maximum operating base-plate temperature,  $T_{\mbox{\tiny B}}$ , is 100 °C. Refer to the thermal derating curve, Figure 5, to see the available output current at baseplate temperatures below 100 °C.

A power derating curve can be calculated for any heatsink that is attached to the base-plate of the converter. It is only necessary to determine the thermal resistance,  $R_{TH_{BA}}$ , of the chosen heatsink between the base-plate and the ambient air for a given airflow rate. This information is usually available from the heatsink vendor. The following formula can then be used to determine the maximum power the converter can dissipate for a given thermal condition:

$$P_{diss}^{max} = \frac{T_{B} - T_{A}}{R_{TH_{DA}}}$$

This value of power dissipation can then be used in conjunction with the data shown in Figure 3 to determine the maximum load current (and power) that the converter can deliver in the given thermal condition.

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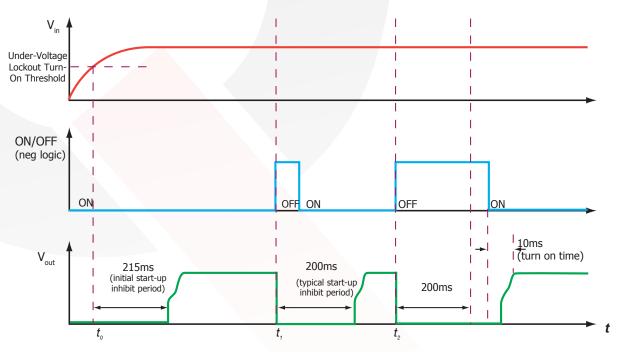
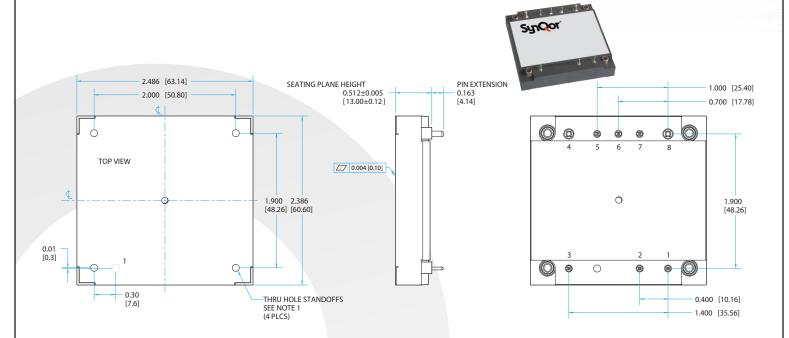


Figure F: Startup Inhibit Period (turn-on time not to scale)



Output: 3.3V

**Current: 60A** 



#### **NOTES**

- THREADED: APPLIED TORQUE PER M3 SCREW 4in-lb RECOMMENDED (5in-lb LIMIT).
   NONTHREADED: DIA 0.125" (3.18mm)
- 2) BASEPLATE FLATNESS TOLERANCE IS 0.004" (.10mm) TIR FOR SURFACE.
- 3) PINS 1-3 AND 5-7 ARE 0.040" (1.02mm) DIA. WITH 0.080" (2.03mm) DIA. STANDOFFS.
- 4) PINS 4 AND 8 ARE 0.080" (2.03mm) DIA. WITH 0.125" (3.18mm) DIA STANDOFFS
- 5) ALL PINS: MATERIAL: COPPER ALLOY FINISH: MATTE TIN OVER NICKEL PLATE
- 6) WEIGHT: 4.7 oz (134 g)
- 7) ALL DIMENSIONS IN INCHES (mm)
  TOLERANCES: X.XXIN +/-0.02 (X.Xmm +/-0.5mm)
  X.XXXIN +/-0.010 (X.XXmm +/-0.25mm)

#### **PIN DESIGNATIONS**

Pin	Name	Function
1	Vin(+)	Positive input voltage
2	ON/OFF	TTL input to turn converter on and off,
		referenced to Vin(-), with internal pull up.
3	IN RTN	Input return
4	OUT RTN	Output return
5	SENSE(-)	Negative remote sense <sup>1</sup>
6	TRIM	Output voltage trim <sup>2</sup>
7	SENSE(+)	Positive remote sense <sup>3</sup>
8	Vout(+)	Positive output voltage

#### Notes

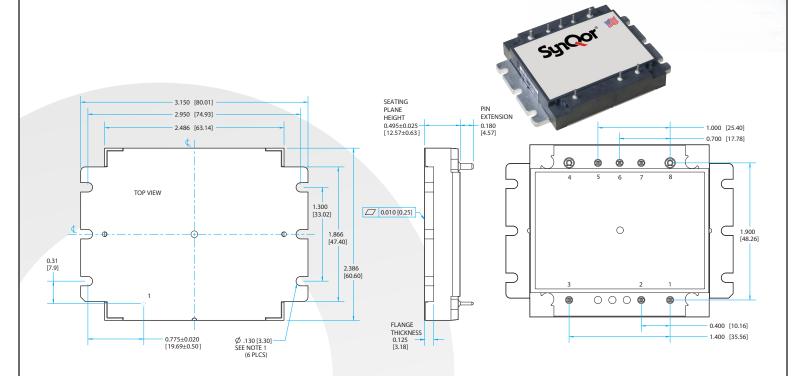
- 1) SENSE(–) should be connected to Vout(–) either remotely or at the converter.
- 2) Leave TRIM pin open for nominal output voltage.
- SENSE(+) should be connected to Vout(+) either remotely or at the converter.

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Output: 3.3V

**Current: 60A** 



#### **NOTES**

- APPLIED TORQUE PER M3 OR 4-40 SCREW 4in-lb RECOMMENDED (5in-lb LIMIT)
- 2) BASEPLATE FLATNESS TOLERANCE IS 0.010" (.25mm) TIR FOR SURFACE.
- 3) PINS 1-3 AND 5-7 ARE 0.040" (1.02mm) DIA. WITH 0.080" (2.03mm) DIA.
- PINS 4 AND 8 ARE 0.080" (2.03mm) DIA. WITH 0.125" (3.18mm) DIA STANDOFFS
- 5) ALL PINS: MATERIAL: COPPER ALLOY FINISH: MATTE TIN OVER NICKEL PLATE
- 6) WEIGHT: 4.9 oz. (140 g)
- 7) ALL DIMENSIONS IN INCHES (mm)
  TOLERANCES: X.XXIN +/-0.02 (X.Xmm +/-0.5mm)
  X.XXXIN +/-0.010 (X.XXmm +/-0.25mm)

#### **PIN DESIGNATIONS**

Pin	Name	Function
1	Vin(+)	Positive input voltage
2	ON/OFF	TTL input to turn converter on and off,
		referenced to Vin(-), with internal pull up.
3	IN RTN	Input return
4	OUT RTN	Output return
5	SENSE(-)	Negative remote sense <sup>1</sup>
6	TRIM	Output voltage trim <sup>2</sup>
7	SENSE(+)	Positive remote sense <sup>3</sup>
8	Vout(+)	Positive output voltage

#### Notes:

- 1) SENSE(-) should be connected to Vout(-) either remotely or at the converter.
- 2) Leave TRIM pin open for nominal output voltage.
- SENSE(+) should be connected to Vout(+) either remotely or at the converter.

Output: 3.3V

**Current: 60A** 

### **Mil-COTS Qualification**

Test Name	Details	# Tested (# Failed)	Consistent with MIL- STD-883F Method	Consistent with MIL-STD- 883F Method 5005	
Visual, mechanical and electrical testing before, during and after 1000 hour burn-in @ full load		15 (0)	Method 1005.8		
Shock-Vibration Visual, mechanical and electrical testing before, during and after shock and vibration tests		5 (0)		MIL-STD-202, Methods 201A & 213B	
<b>Humidity</b> +85°C, 95% RH, 1000 hours, 2 minutes on / 6 hours off		8 (0)	Method 1004.7		
Temperature Cycling	•		Method 1010.8	Condition A	
Solderability 15 pins		15 (0)	Method 2003		
DMT	-65°C to +110°C across full line and load specifications in 5°C steps	7 (0)			
Altitude	70,000 feet (21 km), see Note	2 (0)			

Note: A conductive cooling design is generally needed for high altitude applications because of naturally poor convective cooling at rare atmospheres.

# Mil-COTS DC-DC Converter and Filter Screening

Screening	Process Description	S-Grade	M-Grade	
Baseplate Operating Temperature		-55°C to +100°C	-55°C to +100°C	
Storage Temperature		-65°C to +135°C	-65°C to +135°C	
Pre-Cap Inspection	IPC-610, Class III	•	•	
Temperature Cycling	Method 1010, Condition B, 10 Cycles		•	
Burn-In	100°C Baseplate	12 Hours	96 Hours	
Final Electrical Test	100%	25°C	-55°C, +25°C, +100°C	
Final Visual Inspection MIL-STD-2008		•	•	

# Mil-COTS MIL-STD-810G Qualification Testing

Mil-Co 13 MIL-31D-610G Qualification Testing						
MIL-STD-810G Test	Method	Description				
Fungus	508.6	Table 508.6-I				
Altitude	500.5 - Procedure I	Storage: 70,000ft. / 2 Hr. duration				
Aititude	500.5 - Procedure II	Operating; 70,000ft. / 2 Hr. duration; Ambient Temperature				
Rapid Decompression	500.5 - Procedure III	Storage: 8,000ft. to 40,000ft.				
Acceleration	513.6 - Procedure II	Operating - 15g's				
Salt Fog	509.5	Storage				
High Temperature	501.5 - Procedure I	Storage: 135°C / 3 hrs				
Ingii reinperature	501.5 - Procedure II	Operating: 100°C / 3 hrs				
Low Temperature	502.5 - Procedure I	Storage: -65C / 4 hrs				
Low remperature	502.5 - Procedure II	Operating: -55C / 3 hrs				
Temperature Shock	503.5 - Procedure I - C	Storage: -65C to 135C; 12 cycles				
Rain	506.5 - Procedure I	Wind Blown Rain				
Immersion	512.5 - Procedure I	Non-Operating Non-Operating				
Humidity	507.5 - Procedure II	Aggravated cycle @ 95% RH (Figure 507.5-7 aggravated temp - humidity cycle, 15 cycles)				
Random Vibration	514.6 - Procedure I	10-2000 Hz, PSD level of 1.5 g2/Hz(54.6grms), duration = 1 hr/axis				
Shock	516.6 - Procedure I	20g's peak, 11ms, Functional Shock (Operating no load) (saw tooth)				
SHOCK	516.6 - Procedure VI	Bench Handling Shock				
Sinusoidal vibration	514.6 - Category 14	Rotary wing aircraft - helicopter, 4hrs/axis, 20g's (sine sweep from 10 - 500HZ)				
Sand and Dust	510.5 - Procedure I	Blowing Dust				
Sand and Dust	510.5 - Procedure II	Blowing Sand				

Output: 3.3V

**Current: 60A** 

# **Ordering Information/ Part Numbering**

Example MCOTS-C-48-3R3-HP-N-S

Not all combinations make valid part numbers, please contact SynQor for availability.

Family	Product	Input Voltage	Output Voltage	Package	Thermal Design	Screening Level	Options
MCOTS	C: Converter	<b>28:</b> 16-40V <b>48:</b> 34-75V	1R8: 1.8V 3R3: 3.3V 05: 5.0V 8R4: 8.4V 12: 12V 15: 15V 19R1: 19.1V 24: 24V 28: 28V 40: 40V 48: 48V	<b>HP:</b> Half Brick Peta	N: Normal Threaded D: Normal Non-Threaded F: Flanged	S: S-Grade M: M-Grade	[ ]: Standard Feature

#### **APPLICATION NOTES**

A variety of application notes and technical white papers can be downloaded in PDF format from our website.

#### **PATENTS**

SynQor holds the following U.S. patents, one or more of which apply to each product listed in this document. Additional patent applications may be pending or filed in the future.

5,999,417	6,222,742	6,545,890	6,577,109	6,594,159	6,731,520
6,894,468	6,896,526	6,927,987	7,050,309	7,072,190	7,085,146
7,119,524	7,269,034	7,272,021	7,272,023	7,558,083	7,564,702
7 765 687	7 797 261	8 033 300	8 140 507	Q 403 751	8 644 027

#### Contact SynQor for further information and to order:

Phone: 978-849-0600 Toll Free: 888-567-9596 Fax: 978-849-0602

E-mail: mqnbofae@synqor.com

Web: www.synqor.com Address: 155 Swanson Road

Boxborough, MA 01719

**USA** 

#### Warranty

SynQor offers a two (2) year limited warranty. Complete warranty information is listed on our website or is available upon request from SynQor.

Information furnished by SynQor is believed to be accurate and reliable. However, no responsibility is assumed by SynQor for its use, nor for any infringements of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of SynQor.

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